

Graphene Modulators and Switches Integrated on Silicon and Silicon Nitride Waveguide

Leili Abdollahi Shiramin and Dries Van Thourhout

Abstract—The optoelectronic properties of graphene attracted a lot of interest in recent years. Several demonstrations of integrated graphene based modulators, switches, detectors, and nonlinear devices have been reported. We present here a comprehensive study investigating the different design tradeoffs involved in realizing, in particular, graphene-based modulators and switches. We compare four representative hybrid graphene-waveguide configurations, focusing on optimizing their dimensions, the gate-oxide thickness, the polarization, the operating wavelength, and contact definition. We study both static and dynamic behavior, defining a relevant figure of merit. We find that a 20- μm device based on silicon waveguides should allow for 25-GBit/s modulation rate and an extinction ratio of 5 dB. A 200- μm long SiN device on the other hand should allow for 23-dB extinction ratio and switching speeds down to 0.4 ns.

Index Terms—Double layer, graphene, interconnects, silicon, silicon nitride, waveguide.

I. INTRODUCTION

GRAPHENE has been demonstrated to exhibit unique optoelectronics properties [1]. It has been exploited to demonstrate several photonics devices including photodetectors [2]–[4], modulators [5]–[8], transistors [9], [10], polarizers [11], saturable absorbers [12], [13] and heaters [14].

A promising route to enhance the interaction between light and the graphene layer is to integrate it with an optical waveguide. Integration with different waveguide types, including silicon and silicon nitride waveguides has been demonstrated in literature, but thus far the trade-offs involved in the selection of the optimal waveguide configuration have only be discussed to a limited extent. For example, crystalline silicon waveguides provide more confined modes, possibly leading to more compact devices, while silicon nitride, as a deposited material, is possibly more cost effective and provides more freedom in the positioning of the graphene sheet. The discussed trade-offs are relevant for all devices mentioned above but in this paper we will focus on modulators and switches. For modulators the objective is reaching the highest modulation rate, while maintaining a reasonable extinction ratio (ER) and insertion loss (IL). Both graphene silicon and graphene silicon nitride based modulators

have been proposed in literature. For switches the focus is on obtaining a high ER and low IL, for a switching speed of ~ 1 –10 ns (or slower). Switch architectures proposed in literature so far mostly rely on semiconductor optical amplifiers (SOA) [15], [16], liquid crystals [17], [18] or the thermo-optic effect [19]. The only graphene-based switches demonstrated thus far are a graphene plasmonic mode switch operating at THz frequencies [20] and a device switching the reflectivity for mid-infrared light using back-gated single layer graphene [21]. In this work to the contrary, we focus on the on-off switching of a guided mode in the telecom bands at wavelengths of 1310 and 1550 nm.

The waveguide configurations studied in this work are shown in Fig. 1. The parameters to be optimized include the material platform (Si or Si_3N_4), the polarization of the guided mode, the waveguide dimensions (width and height), the device length and the contact spacing. We focus on non-resonant devices so that we can fully exploit the broadband response of graphene.

II. DEVICE CONFIGURATIONS

The first two configurations are based on a silicon waveguide and were originally proposed in [5] and [6]. The main difference is that the waveguides are assumed to be planarized by chemical mechanical polishing to ease the integration process with the graphene layer (see also [7]). The single layer graphene on silicon configuration (SLG-Si) shown in Fig. 1(a) has a single layer of graphene transferred on the silicon waveguide, with a thin oxide layer in between. The silicon waveguide here serves as the back gate. The double layer graphene on silicon configuration (DLG-Si) has two layers of graphene separated by a thin dielectric, on top [see Fig. 1(b)]. Typically a layer of Aluminum Oxide deposited by atomic layer deposition is used as the dielectric in this capacitive stack. The voltage to drive the device is now applied between top and bottom graphene layer. The double layer graphene on silicon nitride device [DLG-SiN, Fig. 1(c)] is conceptually identical to the DLG-Si, except for the fact that the waveguide is now formed by a lower index Si_3N_4 strip ($n_{\text{Si}_3\text{N}_4} = 2.0$ versus $n_{\text{Si}} = 3.5$). Finally, given that Si_3N_4 is a deposited material, it is possible to embed the double layer graphene stack within the waveguide core, leading to the double layer graphene embedded in SiN device (DLG-E-SiN) shown in Fig. 1(d).

III. STATIC PERFORMANCE

Our simulations were performed based on the surface conductivity method using the finite-element solver COMSOL. The

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The authors are with the Photonics Research Group, INTEC, Ghent University-IMEC, Ghent 9000, Belgium (e-mail: Leili.AbdollahiShiramin@ugent.be; dries.vanthourhout@intec.ugent.be).

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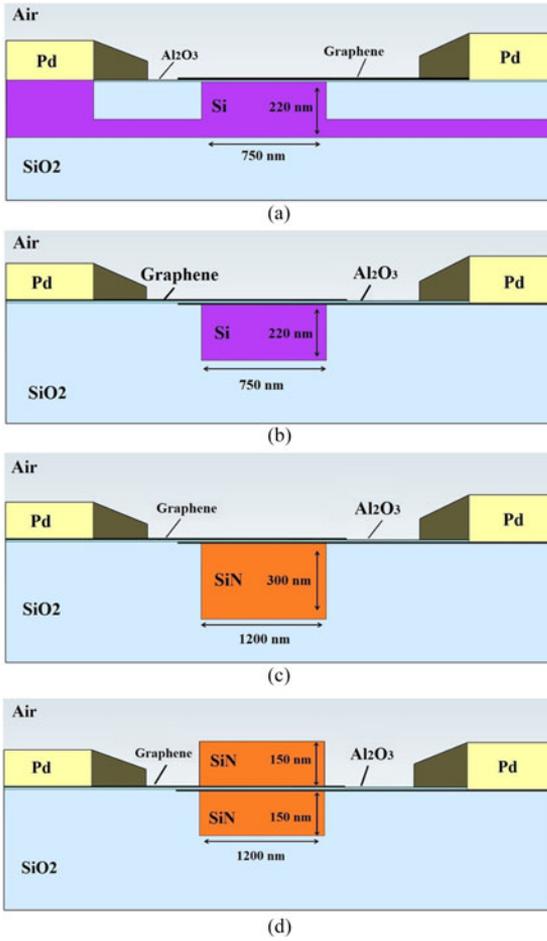


Fig. 1. Graphene based waveguide integrated electro-optical switch. (a) SLG-Si, (b) DLG-Si, (c) DLG-SiN, (d) DLG-E-SiN.

conductivity of graphene was extracted from the Kubo formula [22] taking into account both interband and intraband transitions and depends on the wavelength of the light, the chemical potential μ_c of the graphene layer (controlled through the applied voltage), the carrier relaxation time (τ_r) of the graphene layer (depending on the quality of the graphene after full fabrication) and the temperature of the device. In all simulations we assumed a temperature $T = 300$ K.

Fig. 2 plots the absorption for the different proposed waveguide configurations, as function of the waveguide dimensions, at the neutrality point ($\mu_c = 0$ eV). The absorption is a measure for the interaction between the optical mode and the graphene layer, hence a higher value is preferred. Fig. 2(a) shows that for all structures the absorption increases with increasing waveguide width (assuming waveguide height of 220 nm for the Si structures and 300 nm for the SiN structures). However there are important differences between different configurations. Assuming quasi TM polarization, the DLG-Si devices exhibit almost twice the absorption of the SLG-Si device. For TE-polarization, the absorption in the Si-based devices is considerably lower (up to 0.12 and 0.08 dB/ μ m for DLG-Si and SLG-Si devices, respectively, e.g., see [7], not shown in figure). This is related to the interaction with the strong longitudinal electric field com-

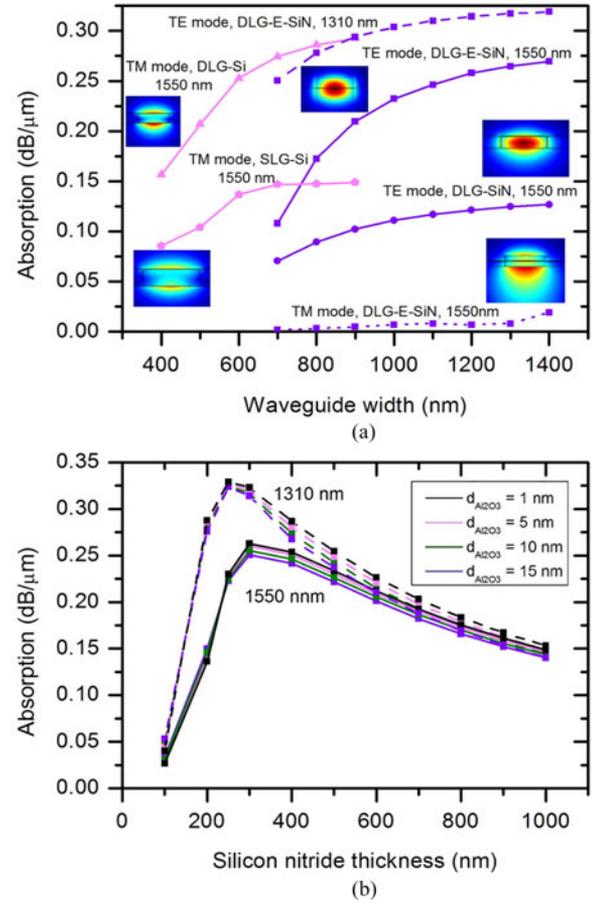


Fig. 2. (a) TE and TM mode absorption in DLG-E-SiN, TE mode absorption in DLG-SiN, TM mode absorption in DLG-Si and SLG-Si as function of waveguide width (SiN: blue, Si: pink), (b) TE mode absorption in DLG-E-SiN as function of bottom SiN thickness. All were simulated for $\tau_r = 65$ fs, $T = 300$, $\lambda = 1550$ nm (and 1310 nm for DLG-E-SiN) while the waveguide width is 1200 nm.

ponent at the top interface of the silicon waveguides (note that the graphene sheet does not interact with the out-of-plane field components [22]). In the SiN-devices on the other hand, the quasi TE-polarized modes exhibit higher interaction with the graphene film than the quasi TM modes. The DLG-SiN has lower absorption than the DLG-Si device but by embedding the graphene stack within the SiN waveguide, the absorption level of DLG-E-SiN can be increased again, almost to the level of the DLG-Si, albeit at larger waveguide dimensions. The simulations shown in Fig. 2(a) were carried out for a gate oxide $d_{Al_2O_3} = 5$ nm. Varying the gate oxide from 1 to 15 nm has very limited impact on the absorption (< 0.03 dB/ μ m).

Finally, it can be seen that, e.g., for the DLG-E-SiN with optimized layer thickness, the absorption at a wavelength of 1310 nm is significantly higher than for 1550 nm. This is related to a higher overlap of the optical mode with the graphene layer at 1310 nm, rather than to a change in the intrinsic absorption properties of graphene, which are fairly constant over the wavelength interval considered.

Fig. 2(b) depicts the effect of the SiN waveguide thickness on the absorption of the DLG-E-SiN device both at a wavelength of

TABLE I
EXTRACTED PARAMETERS FROM FIG. 2 FOR BOTH 1310 AND 1550 NM

Structure	Polarization	Width (nm)	Waveguide thickness (nm)
DLG-E-SiN	TE	1200	300
DLG-SiN	TE	1200	300
DLG-Si	TM	750	220
SLG-Si	TM	750	220

1310 and 1550 nm, assuming the graphene layers are embedded symmetrically in the center of the waveguide. At 1550 nm, the absorption peaks for a total waveguide thickness of 300 nm (150-nm bottom layer, 150-nm top layer), while at 1310 nm it reaches its maximum for a 250-nm thick waveguide. In the remainder of the paper we assume a 300-nm thick waveguide for both wavelengths. Simulations further also showed that the 220-nm thick silicon waveguide widely used in the industry is close to optimal in terms of height.

Table I shows the parameters selected for each configuration and used further in this paper. The widths of 750 and 1200 nm for the Si and SiN waveguides, respectively, are chosen to get high absorption while remaining sufficiently far from the multimode regime.

Applying a voltage over the graphene layer changes its chemical potential μ_c , equivalent to shifting its Fermi level. This implies higher interband transition energy and therefore the absorption of the graphene layer is suppressed. Given the high carrier mobility of graphene ($\sim 10^3 \text{ cm}^2 \text{ V}^{-1} \text{ s}^{-1}$ at room temperature, taking into account a realistic fabrication process [23], [24]) this effect is quasi instantaneous. The change in the absorption as function of voltage is illustrated in Fig. 3 for an Aluminum Oxide layer as a gate oxide with relative permittivity of 5 and thickness of 1, 3 and 5 nm. In calculating these curves we took into account the geometrical capacitance of the stack, using the approach described e.g. in [7], [22] and V_0 which is the offset induced by the background doping n_b of the graphene with $n_b = V_0$ [25], whereby we assumed that both layers are equally but oppositely doped. If this is not the case, e.g. if both layers have the same type of doping, both the maximum achievable absorption and the modulation efficiency will decrease, in worst case to half the level of Fig. 3. The curves are plotted for two values of τ_r , 65 and 13 fs, respectively, whereby the higher τ_r is equivalent with less scattering and hence higher quality graphene (similar values were used in [26]). The change in τ_r has little impact on the maximum absorption but strongly affects the remaining absorption at higher voltages, with strong consequences on the IL and overall figure of merit of the devices as will be discussed further in the paper. The shape of the curves is only determined by μ_c . However, the voltage needed to reach a given μ_c in the graphene layer strongly depends on the gate oxide thickness $d_{Al_2O_3}$ and hence the voltage needed to reach transparency increases for the thicker gate oxides.

From Fig. 3, we can now calculate the modulation efficiency defined as $(\alpha_{90\%} - \alpha_{10\%}) \cdot L / (V\alpha_{10\%} - V\alpha_{90\%})$, the IL and ER for the different devices. The IL is determined at high forward voltage, where the absorption is quasi-minimal while the

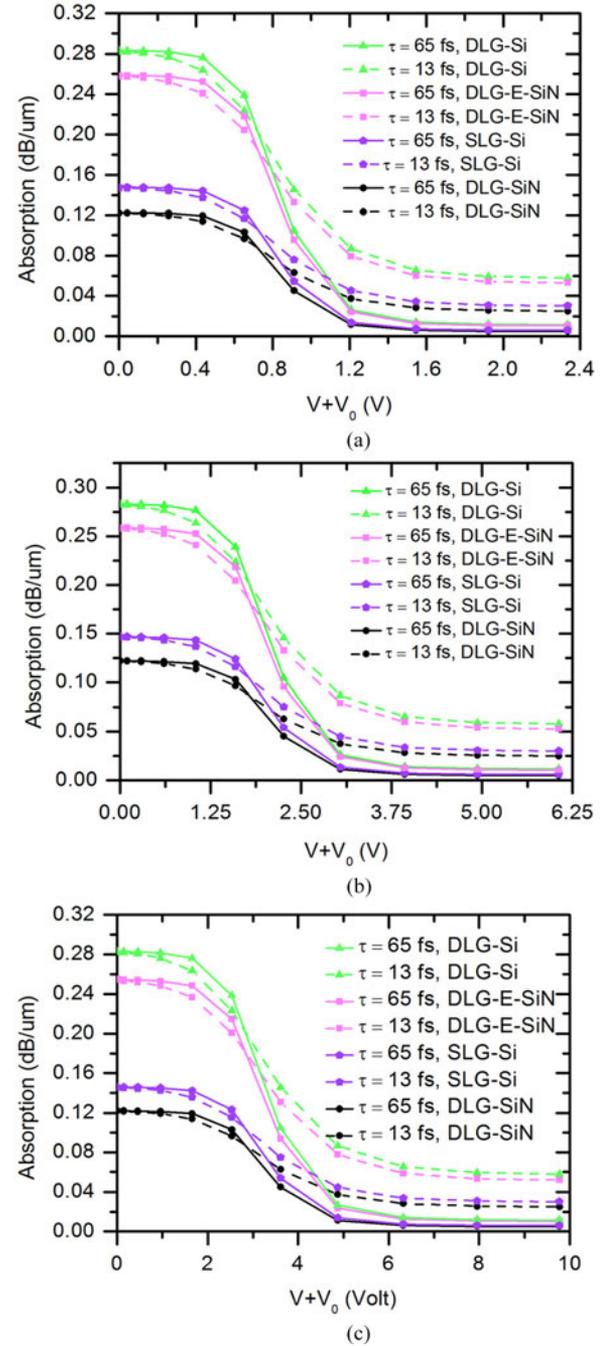


Fig. 3. TE mode absorption in DLG-E-SiN and DLG-SiN and TM mode absorption in DLG-Si and SLG-Si as function of voltage for $\tau_r = 65$ fs and 13 at $\lambda = 1550$ nm with (a) $d_{Al_2O_3} = 1$ nm, (b) $d_{Al_2O_3} = 3$ nm, (c) $d_{Al_2O_3} = 5$ nm.

ER is calculated as the ratio of the maximum and minimum absorption, multiplied by the device length. The results are shown in Fig. 4, as function of length of the device for $\tau_r = 65$ fs.

The modulation efficiency strongly depends on $d_{Al_2O_3}$ and a 5 nm oxide reduces the modulation efficiency considerably in comparison to a 1 nm oxide [see Fig. 4(a)]. IL and ER on the other hand, are nearly independent of $d_{Al_2O_3}$ [see Fig. 4(b)]. Further the modulation efficiency improves for devices, which

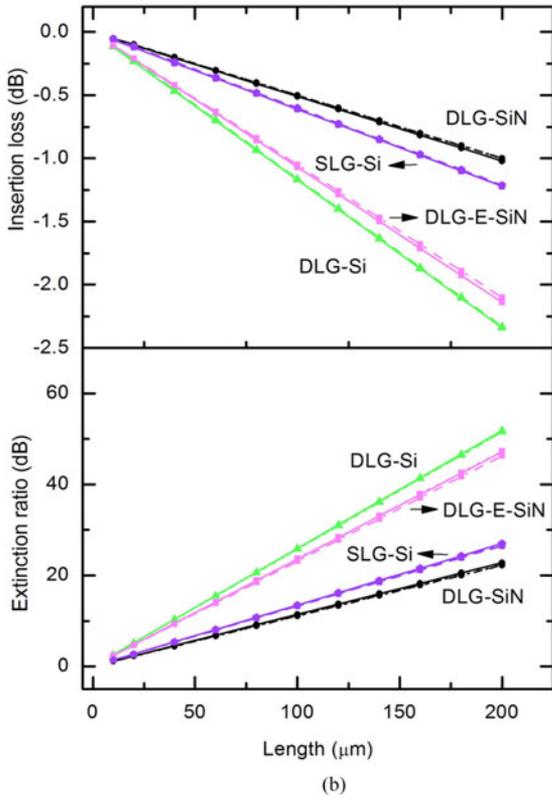
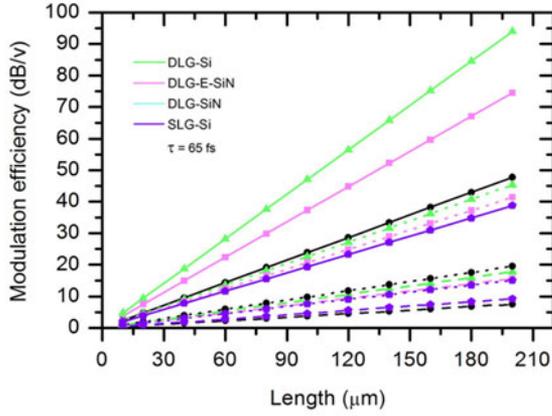


Fig. 4. (a) Modulation efficiency, (b) IL and ER as function of device length for $d_{Al_2O_3} = 1, 3,$ and 5 nm and $\tau_r = 65$ fs. The solid line belongs to oxide thickness of 1 nm, dotted to 3 nm and dashed to 5 nm.

interact stronger with the graphene film (DLG-Si $>$ DLG-E-SiN $>$ SLG-Si $>$ DLG-SiN). This is also true for the ER but the IL shows an opposite behavior: stronger interacting structures exhibit higher excess loss for a given length.

To get more insight in this trade-off, we calculate a figure of merit often used for electro-absorption based modulators, $FOM = ER(\alpha_{V_1} - \alpha_{V_2})/IL(\alpha_{V_2})$ as function of the applied drive voltage ΔV (with $\Delta V = V_1 - V_2$ and $V_{bias} = (V_1 + V_2)/2$ chosen such that the FOM is maximized). The results are shown in Fig. 5. The FOM strongly depends on $d_{Al_2O_3}$, the applied drive voltage and τ_r . Introducing lower quality graphene with shorter scattering time ($\tau_r = 13$ fs vs

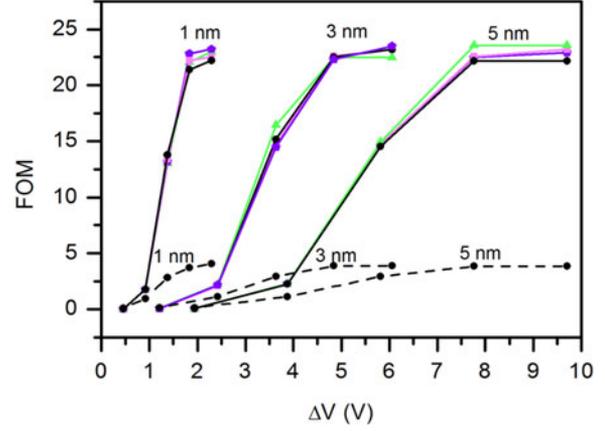


Fig. 5. FOM as function of voltage difference. The solid line is for $\tau_r = 65$ fs, dashed line is for $\tau_r = 13$ fs.

$\tau_r = 65$ fs) reduces the best achievable FOM by almost a factor of five. But, interestingly, the FOM is independent of the chosen device configuration. A similar conclusion was reached in a more general context in [27]. Hence, only considering the static performance of the device one should select the thinnest oxide layer technologically feasible while the detailed device configuration can be decided on other grounds, e.g. compatibility with other devices or cost. This conclusion will no longer be correct however when taking also the dynamic response into account as discussed in the next section.

Thus far we only considered the IL related to the graphene layer itself. However, we also have to take into account the losses caused by the metal contacts and by the resistive part of the graphene outside of the capacitive stack [see Fig. 6(a)]. Here we have assumed Palladium as the contact because of its low contact resistance on graphene [28]–[30]. Fig. 6(b) shows the impact of the metal contacts as function of its spacing (s) with the waveguide, assuming, for technological reasons, that the graphene capacitive stack itself extends 300 nm besides the waveguide ($d_O = 300$ nm). Extending the spacing beyond $0.75 \mu\text{m}$ (DLG-Si) and $1 \mu\text{m}$ (others) is sufficient to suppress the loss of the metal contacts almost completely. The remaining loss then stems from the overlap with the non-transparent graphene (absorber graphene, d_A) outside the capacitive stack. Fig. 6(c) shows the IL as function of the parameter d_O , which indicates how far the two overlapping graphene layers extend besides the waveguide. Increasing d_O reduces the loss due to the non-transparent graphene. However, note that this also has a strong impact on the overall capacitance of the device, negatively impacting its modulation speed. Similarly increasing the metal contact spacing s [see Fig. 6(b)] increases the series resistance of the device again resulting in a reduced modulation speed. Therefore, as will be shown in the next section, in particular for high speed modulators it might be beneficial to tolerate some added IL and reduce both d_O and d_A to maximize the speed.

All the devices discussed above operate over a wide wavelength range, inheriting the broadband modulation properties of graphene. As an example Fig. 7 shows the modulation curves for

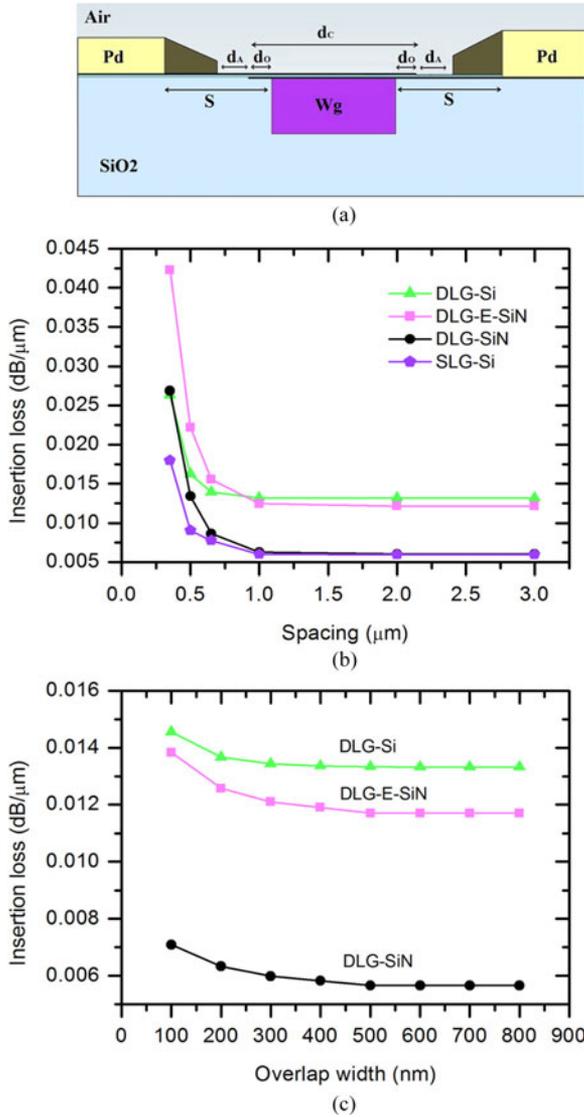


Fig. 6. (a) Schematic figure describing the parameters used in calculating the metal IL, (b) IL as function of metal and waveguide spacing s for $d_{Al_2O_3} = 3$ nm with $\tau_r = 65$ fs and overlap width d_o of 300 nm, (c) IL as function of overlap width d_o for spacing of 1 μ m.

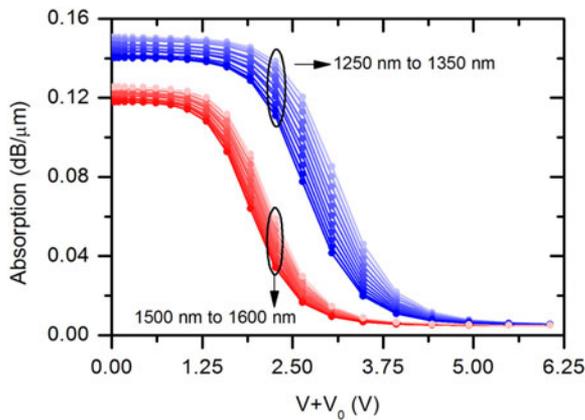


Fig. 7. TE mode absorption of DLG-SiN as function of voltage for $d_{Al_2O_3} = 3$ nm and with the $\tau_r = 65$ fs for the wavelength of 1500–1600 nm and 1250–1350 nm.

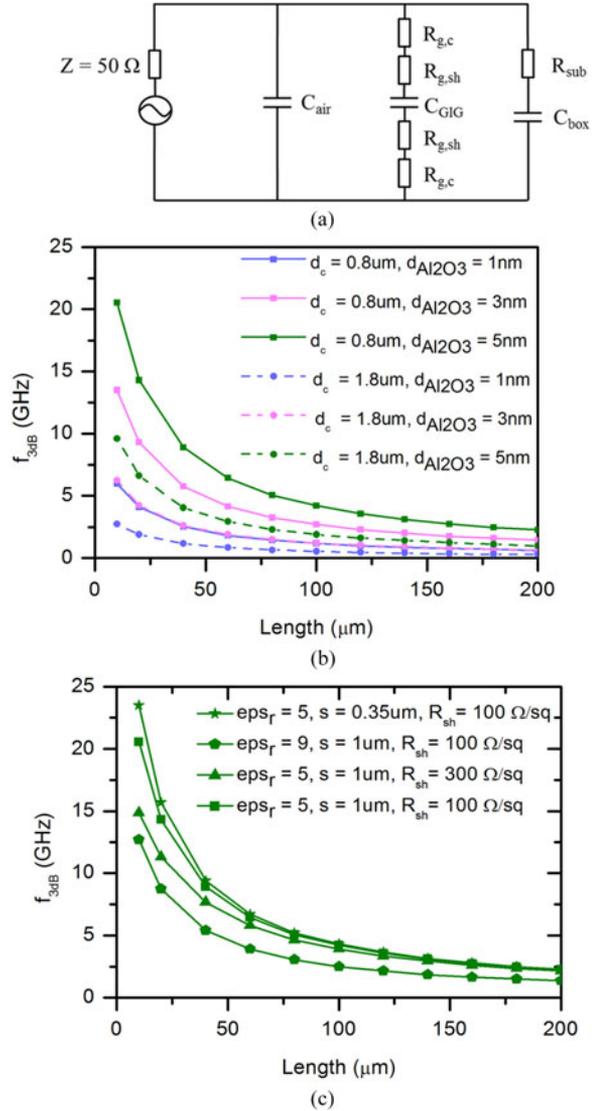


Fig. 8. (a) Equivalent electrical circuit, (b) electrical 3-dB bandwidth as function of device length for $d_{Al_2O_3} = 1, 3,$ and 5 nm. All for $\epsilon_{ps_r} = 5, s = 1 \mu m, R_{sh} = 100 \Omega/sq$, (c) electrical 3-dB bandwidth as function of device length for different relative permittivity, spacing, and graphene sheet resistance all for $d_c = 0.8 \mu m$ and oxide thickness of 5 nm.

a DLG-SiN device ($d_{Al_2O_3} = 3$ nm, $\tau_r = 65$ fs). Nearly wavelength independent operation is obtained from 1500 to 1600 nm and from 1250 to 1350 nm. Operating at 1310 nm results in a higher absorption as discussed above but also requires a higher drive voltage.

IV. DYNAMIC PERFORMANCE

The frequency response of the devices can be calculated using the electrical circuit model shown in Fig. 8(a), similar to the one used in ref [26]. R_{sub} and C_{box} present the substrate resistance and buried oxide capacitance, C_{air} is the capacitance between the electrodes through the air and C_{GIG} is the capacitance of the graphene-insulator-graphene stack. The series resistance, R_s , is the sum of the contact resistance ($R_{g,c}$) and sheet resistance ($R_{g,sh}$) of the first and second graphene layer.

C_{GIG} comprises the oxide capacitance (C_{ox}) and quantum capacitance (C_q) of top and bottom graphene layers [31] where C_{ox} was calculated using a simple parallel plate approximation and C_q is dependent on the carrier density in the graphene sheets. This means C_q and hence the modulation bandwidth f_{3dB} are voltage dependent. Therefore we calculated f_{3dB} at the bias voltage exhibiting the highest modulation efficiency (1.6, 4.27 and 6.9 V for 1, 3 and 5 nm oxide thickness, respectively). C_q was calculated assuming a background doping of $n_b = 9 \times 10^{12} \text{ cm}^{-2}$ and the doping from bias voltage. The relative permittivity of Aluminum Oxide varies for different fabrication processes, hence in our calculations it is assumed to be 5 or 9. The graphene contact resistance is assumed to be $R_{g,c} = 150 \Omega \cdot \mu\text{m}$, a value achievable e.g. through edge contacting of graphene [23], [24], [27]. The graphene sheet resistance $R_{g,sh}$ is varied between 100 and 300 Ω/sq , equivalent with a mobility of 6900 and 2300 $\text{cm}^2/\text{V}\cdot\text{s}$, respectively, assuming the same background doping as used for the C_q calculation ($n_b = 9 \times 10^{12} \text{ cm}^{-2}$) and the spacing s between the waveguide edge and the metal contact edge was varied between 1 and 2 μm . The effect of d_O is included in the capacitance width (d_C), which comprises the waveguide width and twice the overlap region d_O . C_{air} , C_{box} and R_{sub} were taken from [32].

The electrical bandwidth was calculated assuming a 50 Ω source resistance [see Fig. 8(b)–(c)]. Fig. 8(b) shows how the modulation bandwidth changes as function of device length for varying oxide thickness and capacitor width (d_C). The fact that the 50 Ω source resistance is of the same order as the intrinsic device resistance comprising $R_{g,c}$ and $R_{g,sh}$ and varies roughly between 1.85 Ω and 90 Ω for the parameters considered in Fig. 8(b)–(c), makes that the overall modulation bandwidth strongly decreases with increasing length. Further one notes that a smaller d_C and a thicker oxide lead to higher bandwidth. However, as discussed in the previous paragraph, these also lead to lower modulation efficiency and higher IL . A 200 μm long device with $d_C = 1.8 \mu\text{m}$, $d_{Al_2O_3} = 3 \text{ nm}$ has a bandwidth 2.5 GHz, compatible with 0.4 ns switching time. Assuming a DLG-SiN configuration, and a 3 V drive voltage this device also exhibits a low IL (1dB) and a high ER (23 dB). Alternatively, a 20 μm long DLG-Si device with $d_C = 0.8 \mu\text{m}$ and $s = 0.35 \mu\text{m}$, $d_{Al_2O_3} = 5 \text{ nm}$ is compatible with 16 GHz modulation (shown in Fig. 8(c)) and exhibits reduced IL (0.25 dB) but also a reduced ER (5.2 dB).

Fig. 8(c) also shows the effect of the sheet resistance $R_{g,sh}$ and the relative permittivity of the oxide layer. Obviously a lower sheet resistance and reduced spacing s are beneficial for the performance of the device. Increasing the relative permittivity of the oxide layer on the other hand is equivalent to decreasing its speed.

V. CONCLUSION

In summary, we investigated the performance of four representative configurations, relying either on Silicon or on Silicon Nitride waveguides, for realizing graphene based switches and modulators operating in the wavelength regions around 1310 and 1550 nm. We first studied their properties at static operating conditions, which might actually also be relevant for other

types of devices such as voltage controlled non-linear devices and detectors. Interestingly we found that the figure-of-merit of these devices, defined as the ratio of the ER over the IL , is independent of the precise waveguide configuration and only depends on the quality of the used graphene and the gate-oxide thickness. Hence the waveguide configuration can be selected based on other criteria. For ultimate speed, the high confinement offered by higher index contrast silicon waveguides offers more compact and hence faster devices. For switches or non-linear devices with voltage controlled doping, which have less stringent requirements on operating speed, deposited silicon nitride waveguides can provide more versatility and do not suffer from two photon absorption. In all applications accurately controlling the gate-oxide thickness is extremely important as it directly impacts modulation efficiency, power consumption and operation bandwidth.

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Dries Van Thourhout received the Master's degree in physical engineering and the Ph.D. degree from Ghent University, Ghent, Belgium, in 1995 and 2000, respectively.

From October 2000 to September 2002, he was with Lucent Technologies, Bell Laboratories, New Jersey, USA, working on the design, processing, and characterization of InP/InGaAsP monolithically integrated devices. In October 2002, he joined the Department of Information Technology, Ghent University, Belgium, where he is currently a Full Professor. His research focuses on the design, fabrication, and characterization of integrated photonic devices. Main topics involve silicon nanophotonic devices and the integration of novel materials (graphene, III–V, ferro-electrics, quantum dots, etc.) on the waveguides to expand their functionality. He is working on applications for telecom, diatom, optical interconnect, and sensing.



Leili Abdollahi Shiramin received the B.Sc. and M.Sc. degrees in physics from the University of Tabriz and K. N. Toosi University of technology, Tehran, Iran, in 2008 and 2011, respectively. From 2011 to 2013, she joined to the Photonics Group at University of Tabriz as a Researcher. Since 2013, she has been a Ph.D. Researcher at Photonics Research Group, Ghent University-Imec, Ghent, Belgium. Her current interest includes 2-D material-based optical devices fabrication and characterization. She is a Student Member of the IEEE Photonics Society.